

Lecture 5

Digital television

The forward error correction (FEC)

- Need for error correction
- Overall picture
- The different FEC methods in DVB:
 - PRBS energy dispersal (bit level coding)
 - Reed Solomon outer coder (byte/block level)
 - Outer interleaver (byte level)
 - Convolution inner coder (Viterbi decoder) (bit level)

DVB forward error correction

The Transport stream should be transmitted over a Quasi Error Free (QEF) channel

-> (Bit Error Ratio) $BER < 10^{-10}$

1 error on 10^{10} bits

Compare:

1. TCP – Error free channel by resending packes

Resending the date 3 times and voting on the correct?

Basis for error correction

Hamming distance of x and y

$$d(x, y) := |\{i \mid 1 \leq i \leq n, x_i \neq y_i\}|$$

i.e. the number of bits that are different in x and y .

$$x = 0111\mathbf{000}1110$$

$$y = 0111\mathbf{101}1110$$

$$d(x, y) = 2$$

Hence, we try to find a code C that maximizes the distance between any two codewords in C

Example

Parity check

Parity check bit added to codeword

0111 **1**

0101 **0**

The minimum hamming distance between any two codeword in a parity check code is 2.

Hence, the codeword length is 5 bits instead of 4 bits, but a single bit error can now be detected.

Error correction

Parity checking enables error detection. It is also possible to construct codes, that can perform error correction.

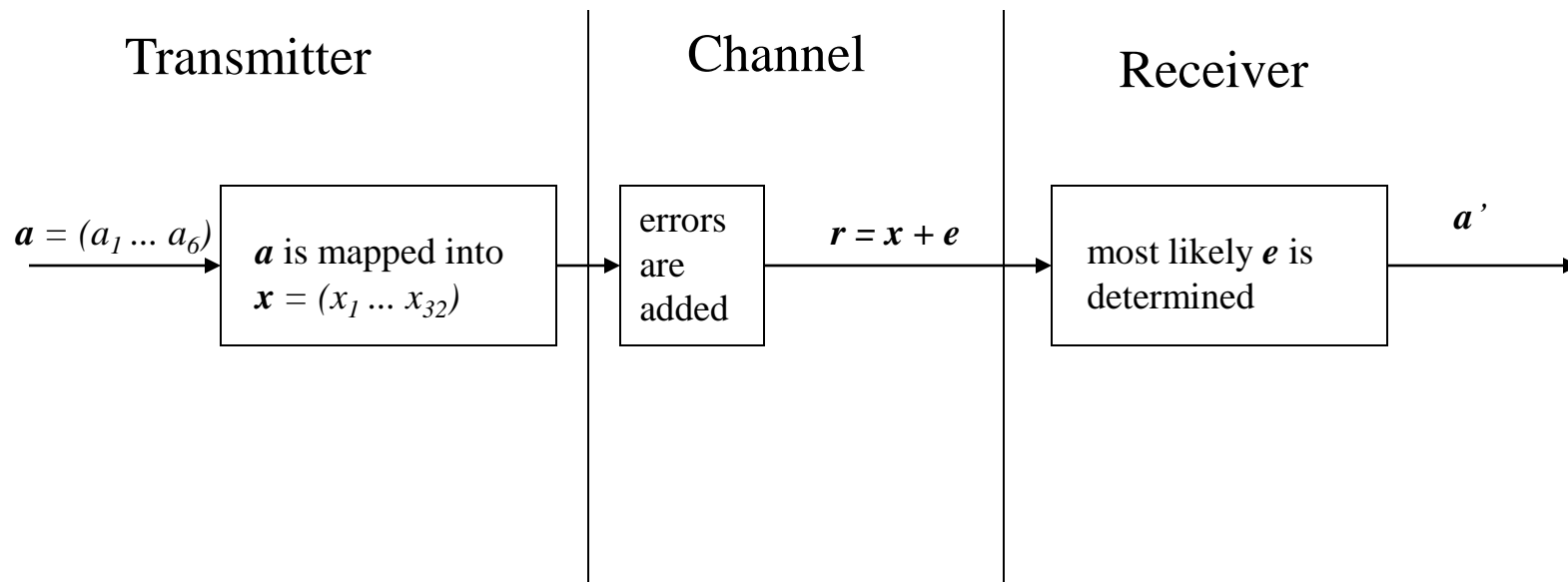
Idea: construct a code, where the Hamming distance between any two codewords is as large as possible.

For any received word r , compare the received word with all codewords in the code, select the codeword x where

$$d(x^i, r) < a$$

Code		Distances	Decoded
01001	Received	2	
01110	01111	1	01110
10011		3	
11000		4	

Forward error correction



Coding gain

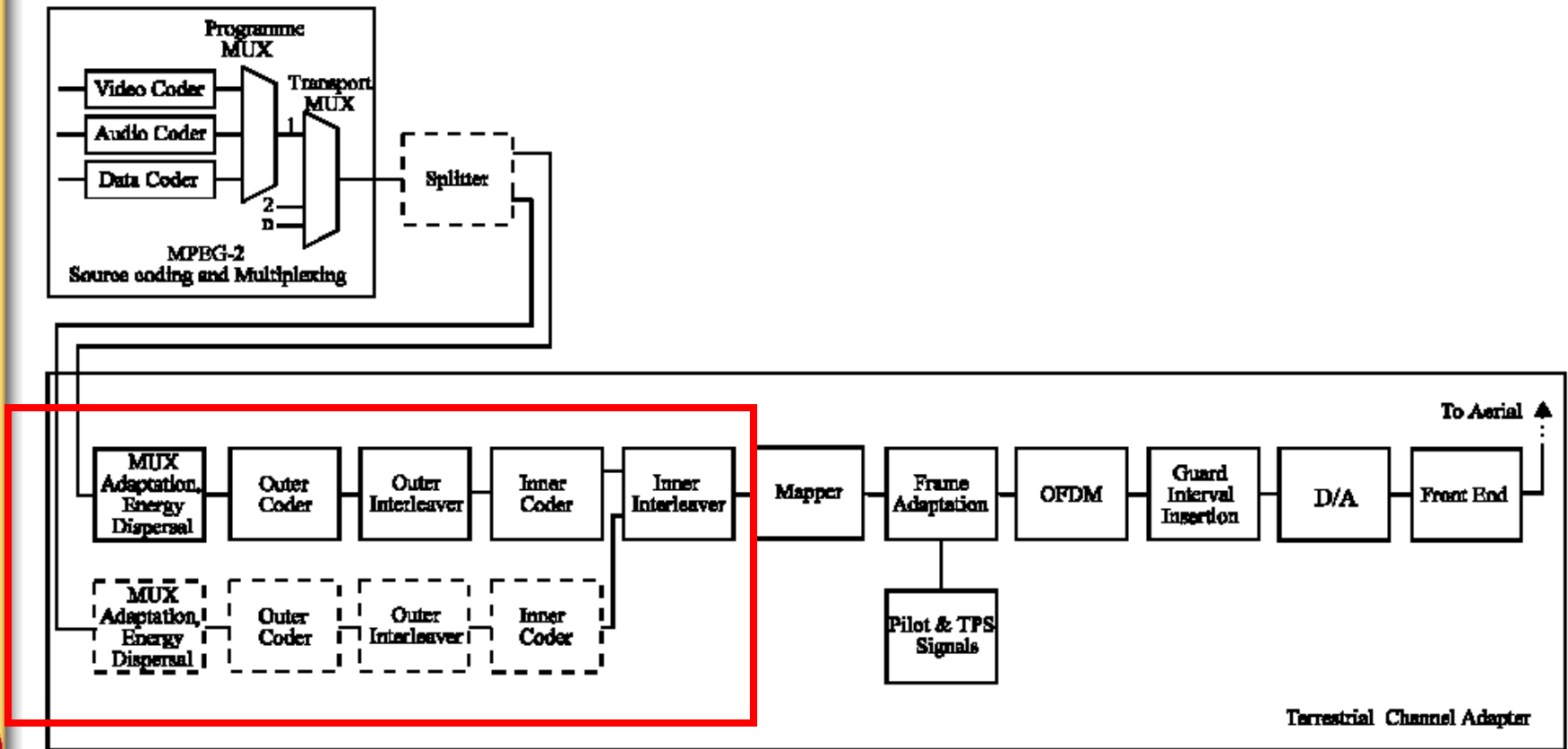
B – number of bits per second we would like to transmit

W – power used for transmitting in noisy channel

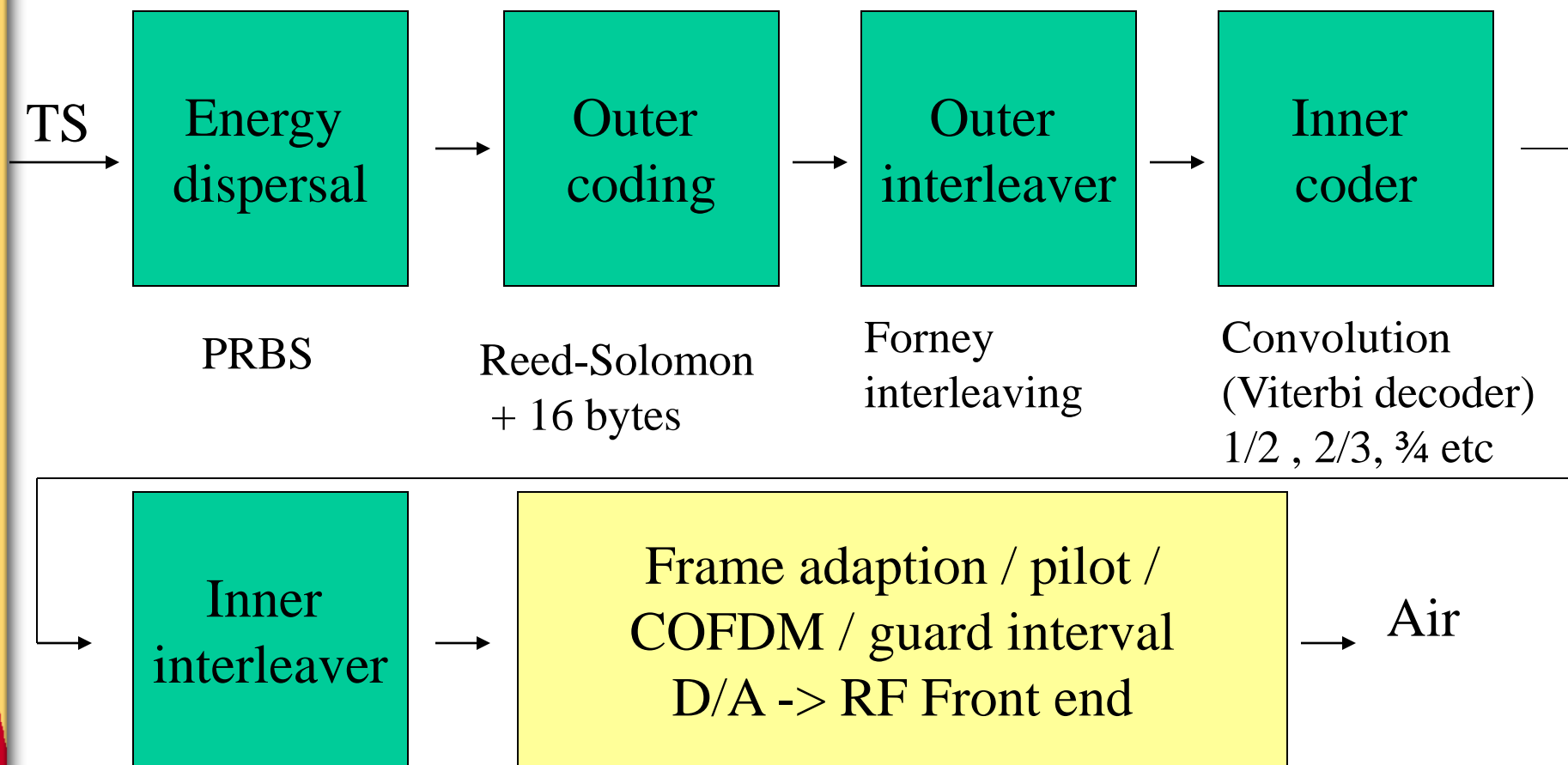
R – code rate = k/n k-original word length, n-word length

Coding only makes sense if the error rate using coding is less than without coding = *coding gain*

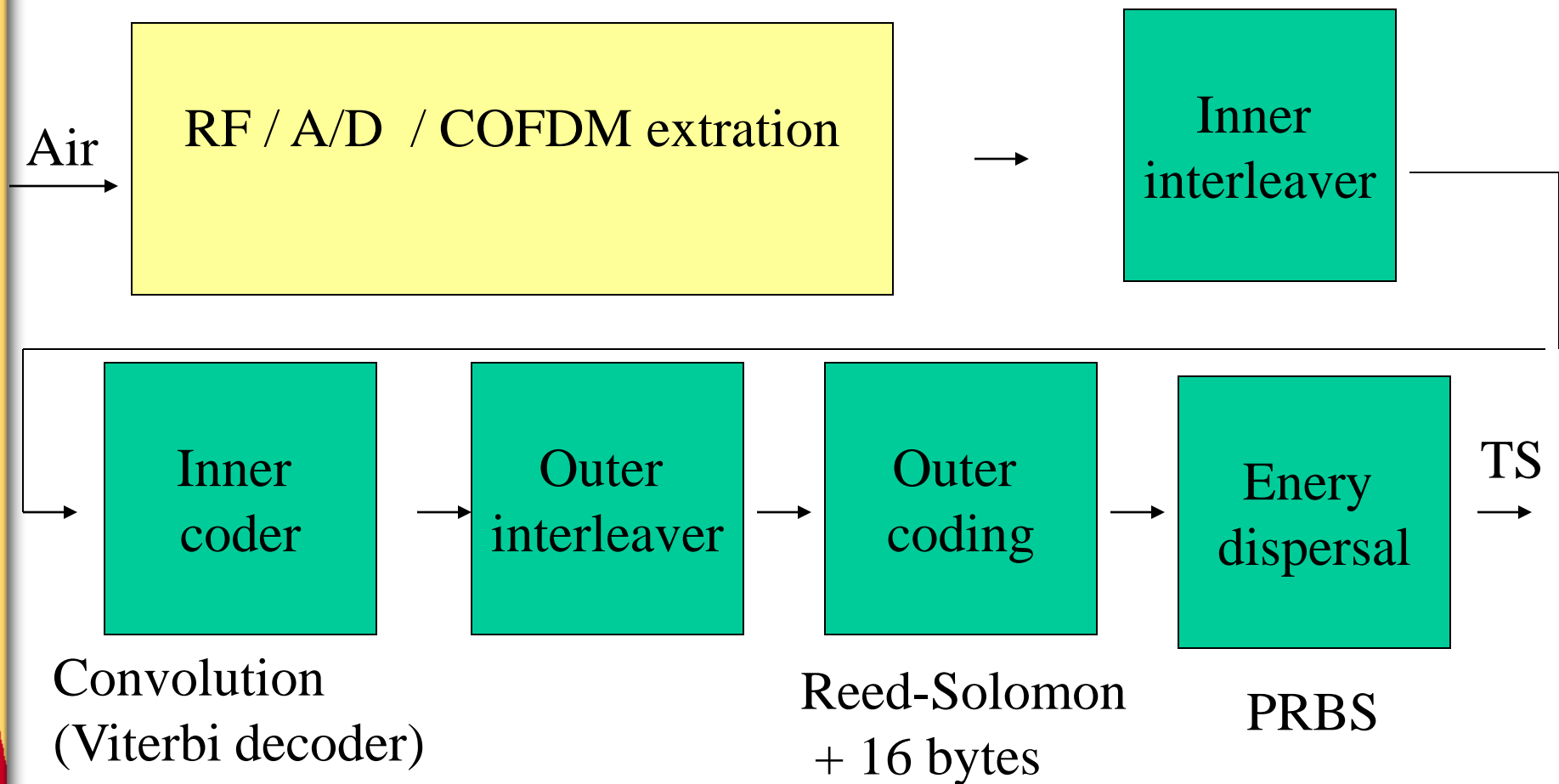
DVB overall transmission sequence



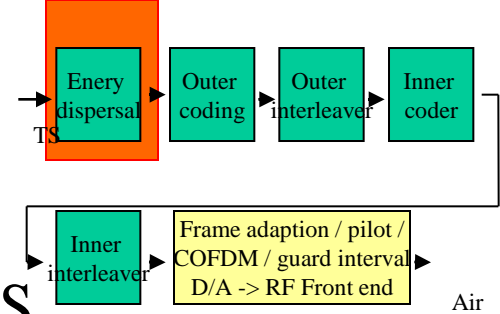
FEC overall transmission



FEC receiver end



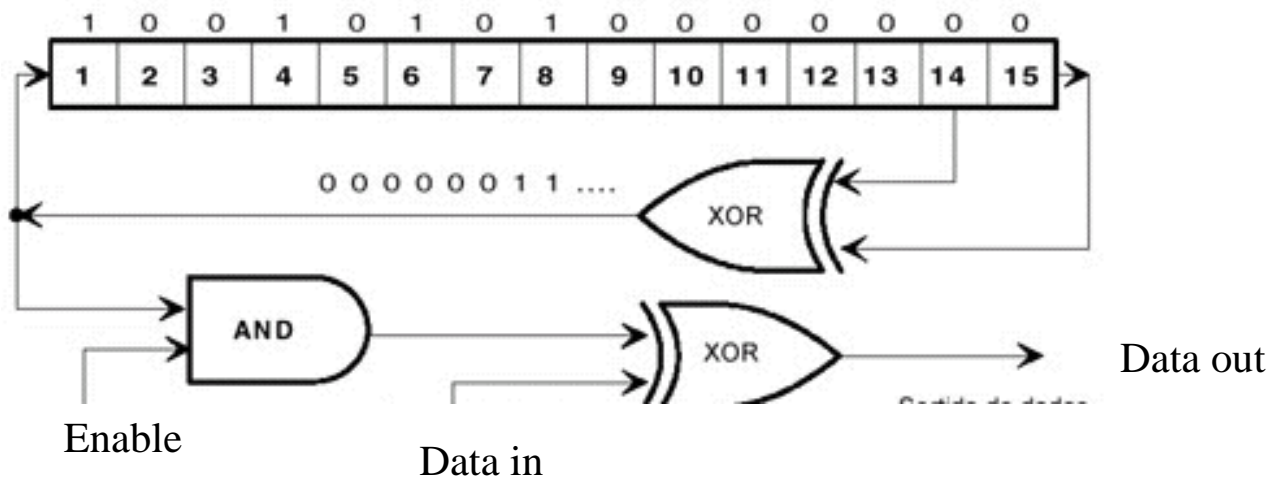
Energy dispersal



Pseudo Random Binary Sequence – PRBS

Polynomial generating the sequence:

$$1 + X^{14} + X^{15}$$



Energy dispersal example

Test: 1
1
1
1 1 1 1 1 1 1

PRBS: 0 0 0 0 0 0 1 0 1 0 1 0 0 1 1 1 1 1 1 1 0 1 0 1 0 1 1 0
0 0 0 0 0 0 1 0 1 0 1 0 0 1 1 1 1 1 1 1 0 1 0 1 0 1 1 0 0 0 0 0
0 0 1 0 1 0 1 0 0 1 1 1 1 1 1 1 0 1 0 1 0 1 1 0 0 0 0 0 0 0 1 0
1 0 1 0 0 1 1 1

Reed solomon

Block level code (works on bytes)

- Adds extra bytes for error correction
- Need for a block synchronization

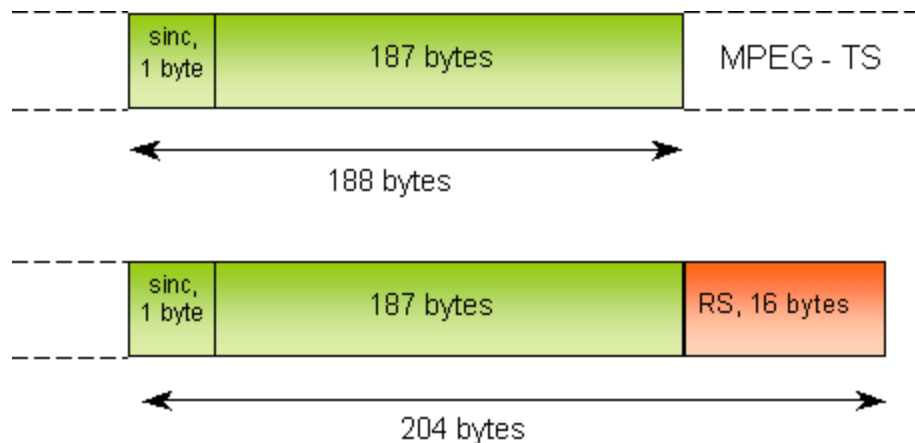
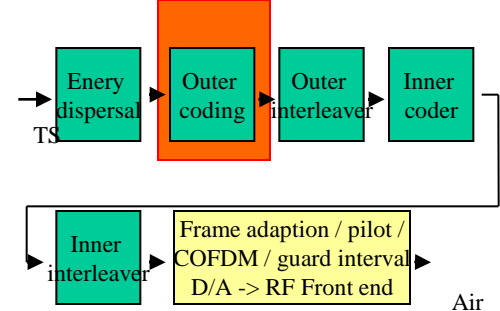
Characteristics: n, k, t

n = 204 Final transport packet length

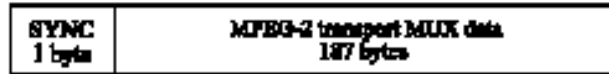
k = 188 Original correctable bytes

t = 8 Number of correctable bytes

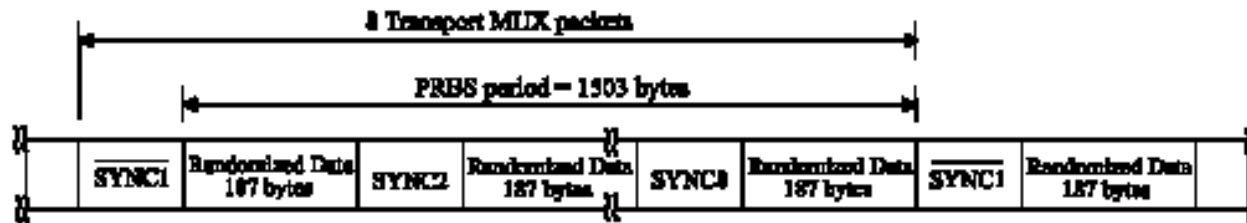
-> RS(204, 188)



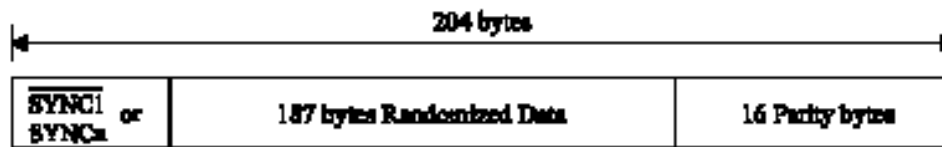
Reed Solomon



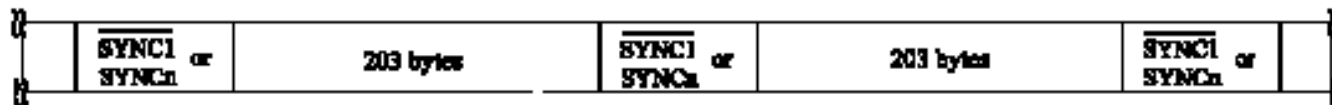
a) MPEG-2 transport MUX packet



b) Randomized transport packets: Sync bytes and Randomized Data bytes



c) Reed-Solomon RS(204,188,8) error protected packets



d) Data structure after outer interleaving; interleaving depth $I = 12$ bytes

Reed Solomon mathematics

Finite (Galois) Field Arithmetic

- Reed-Solomon codes are based on a specialist area of mathematics known as Galois fields or finite fields.
- A finite field has the property that arithmetic operations (+, -, \times , / etc.) on field elements always have a result in the field.
- A Reed-Solomon encoder or decoder needs to carry out these arithmetic operations. These operations require special hardware or software functions to implement.

Reed Solomon if error?

When a codeword is decoded, there are three possible outcomes:

1. If $2s + r < 2t$ (s errors, r erasures) then the original transmitted code word **will always be recovered**,

OTHERWISE

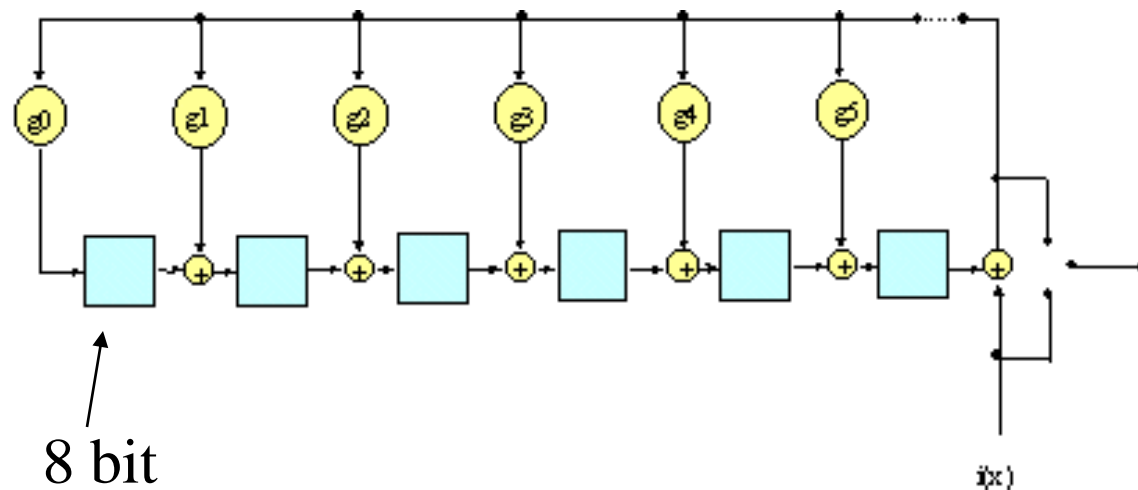
2. The decoder will detect that it cannot recover the original code word and indicate this fact.

OR

3. The decoder will mis-decode and recover an incorrect code word without any indication.

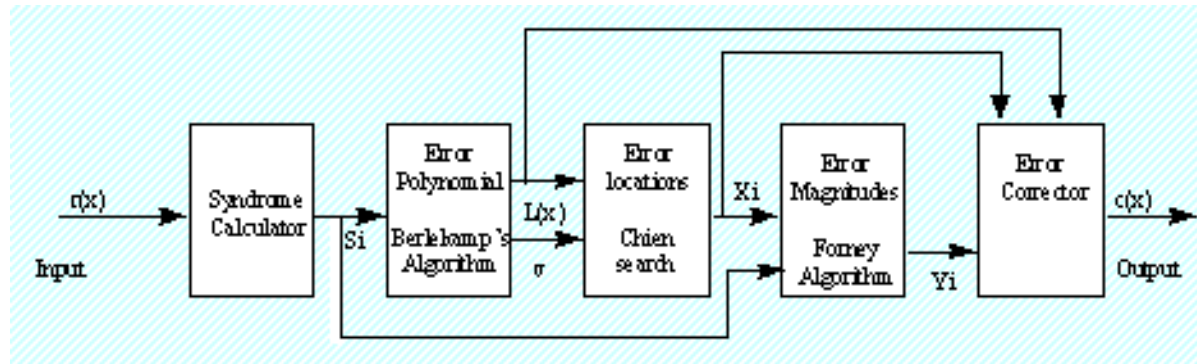
The probability of each of the three possibilities depends on the particular Reed-Solomon code and on the number and distribution of errors.

Reed Solomon encoder architecture



Finite field arithmetics

Reed Solomon decoder architecture



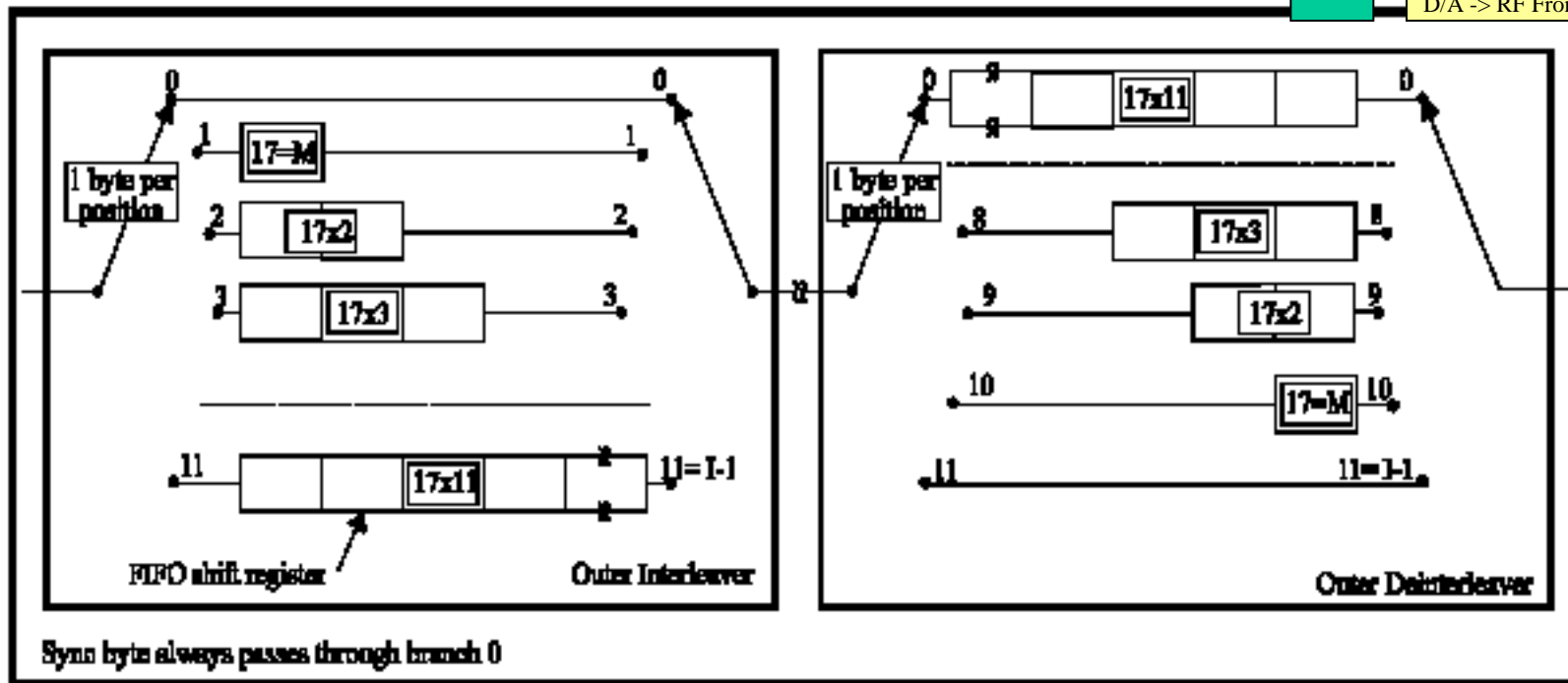
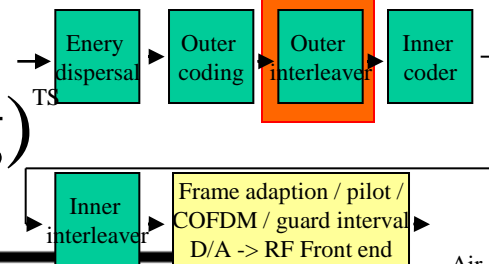
- $r(x)$ Received codeword
- S_i Syndromes
- $L(x)$ Error locator polynomial
- X_i Error locations
- Y_i Error magnitudes
- $c(x)$ Recovered code word
- v Number of errors

Reed Solomon in software

166 Mhz Pentium

Code	Data rate
RS(255,251)	12 Mbps
RS(255,239)	2.7 Mbps
RS(255,223)	1.1 Mbps

Outer interleaving (and deinterleaving)



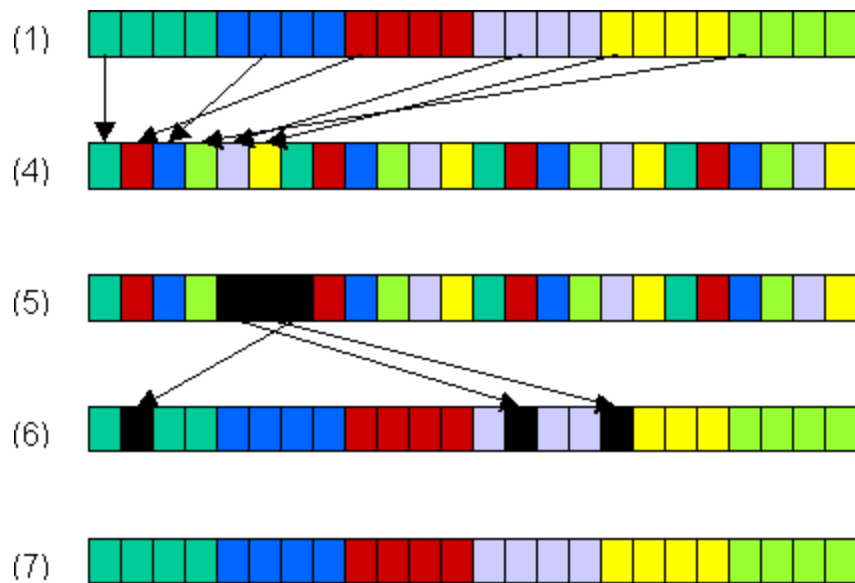
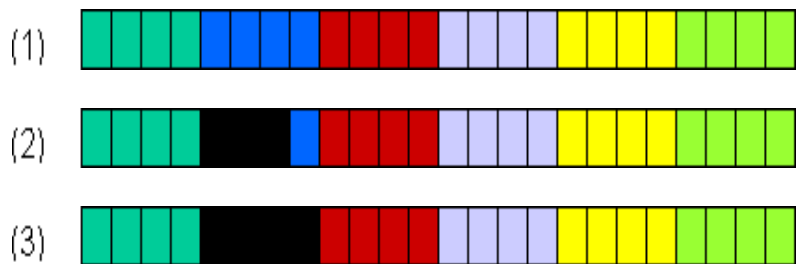
In DVB: Number of branches $i = 12, \{0, \dots, 11\}$

Length of packet to be protected $L = 204$

$M = (L/i) * j (=17*j), j$ branch number index

Interleaving

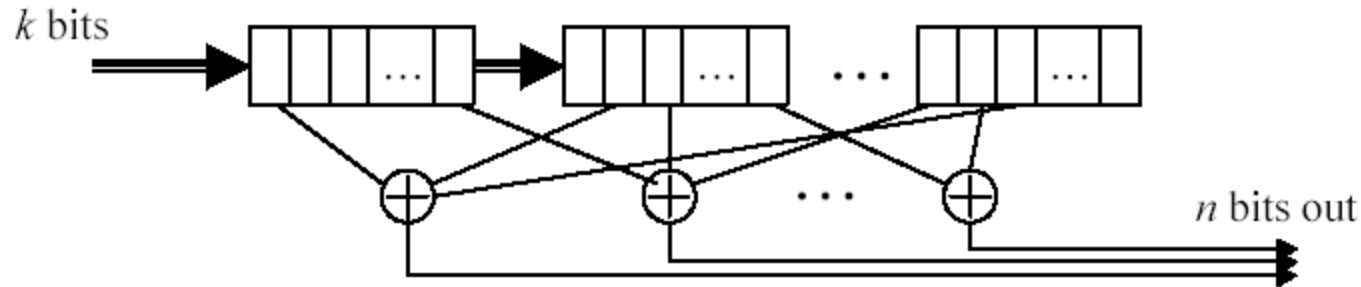
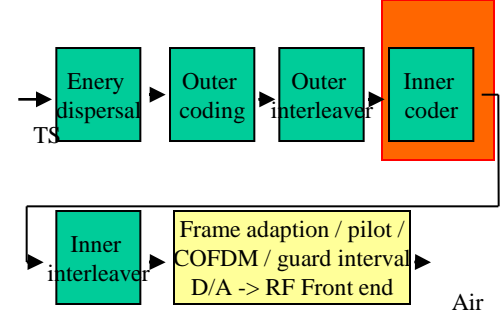
No interleaving



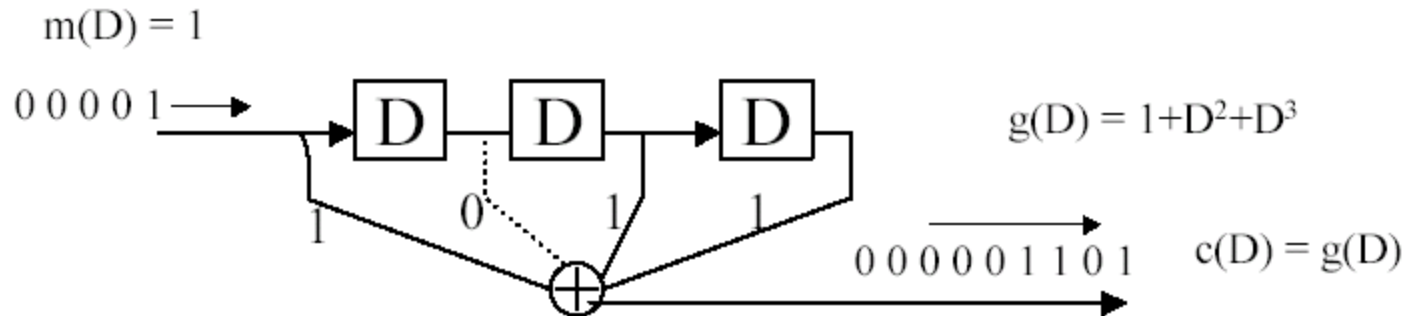
Viterbi + Reed Solomon

Convolution Codes

- Code words are generated by $c(D) = m(D)g(D)$
 - D is one unit delay of a shift register circuit
 - $g(D)$ is realized with a linear finite-state shift register
 - The degree of $m(D)$ can be infinite. So can be that of $c(D)$.
- Rate k/n code,
 - k information bits get shifted in at each D ,
 - goes thru K units of delay, (K is called the *constraint length*)
 - generates n coded bits output at each delay

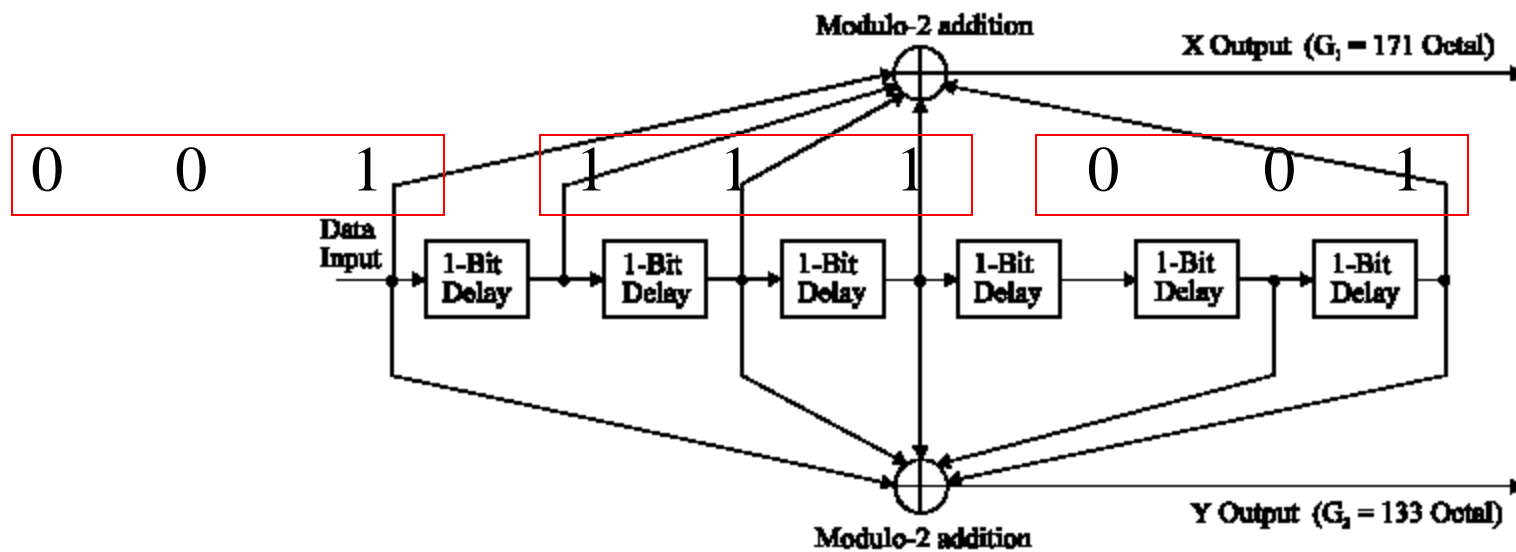


Convolution



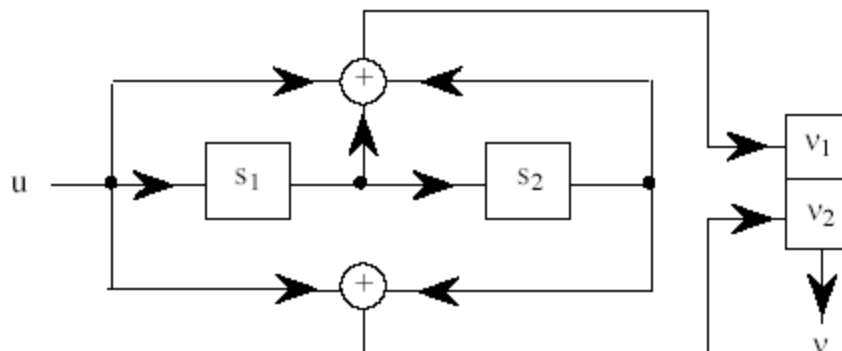
- The convolution is
 - Shifting the input sequence
 - Multiplying the shifted input term by term with the filter coefficients
 - Add up the product terms
 - Shift the input sequence by one and repeat
- Here the input and the filter coefficients are binary and the summation is module 2 addition

Convolution coding in DVB

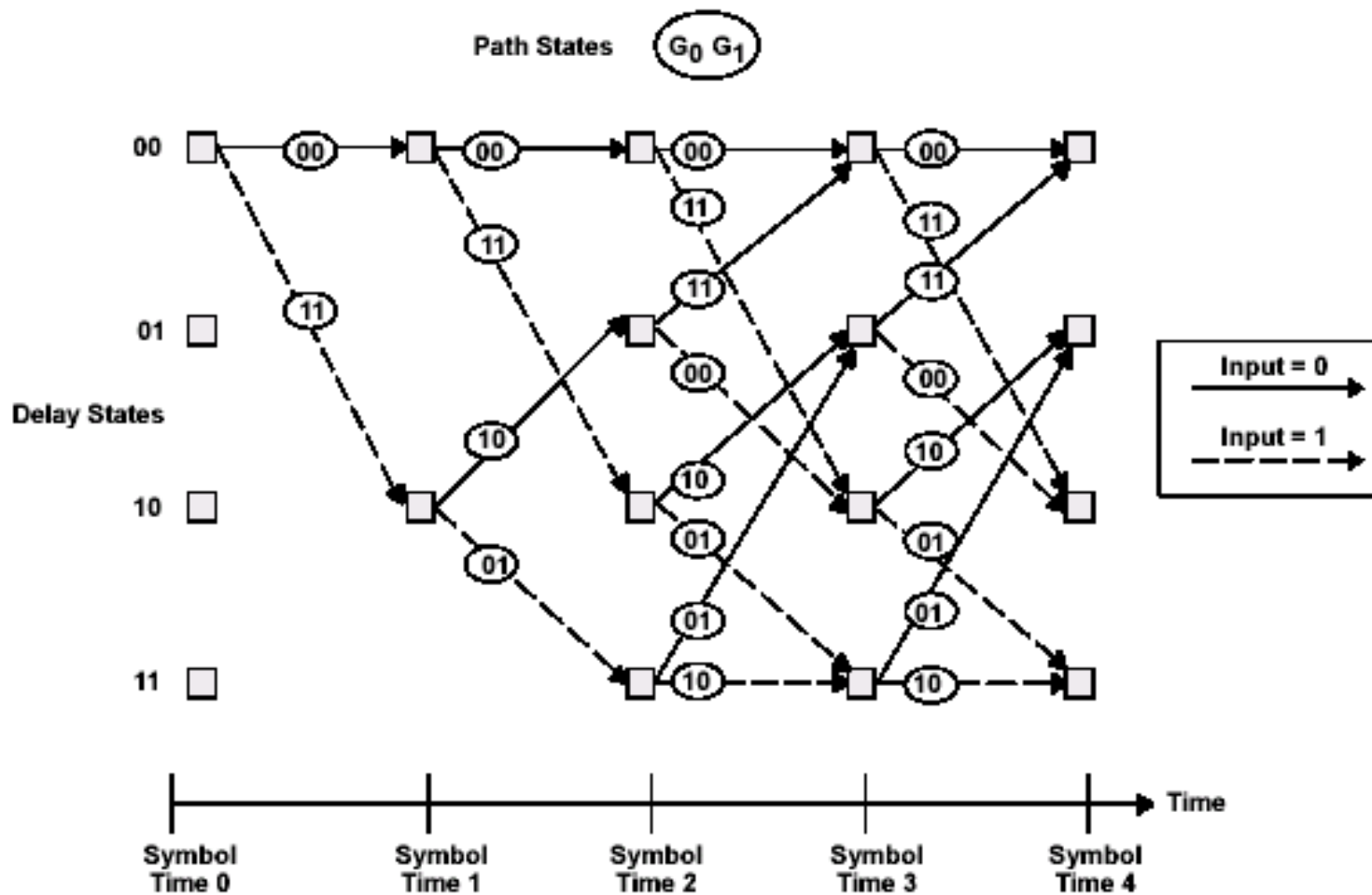


Decoding the convolution code

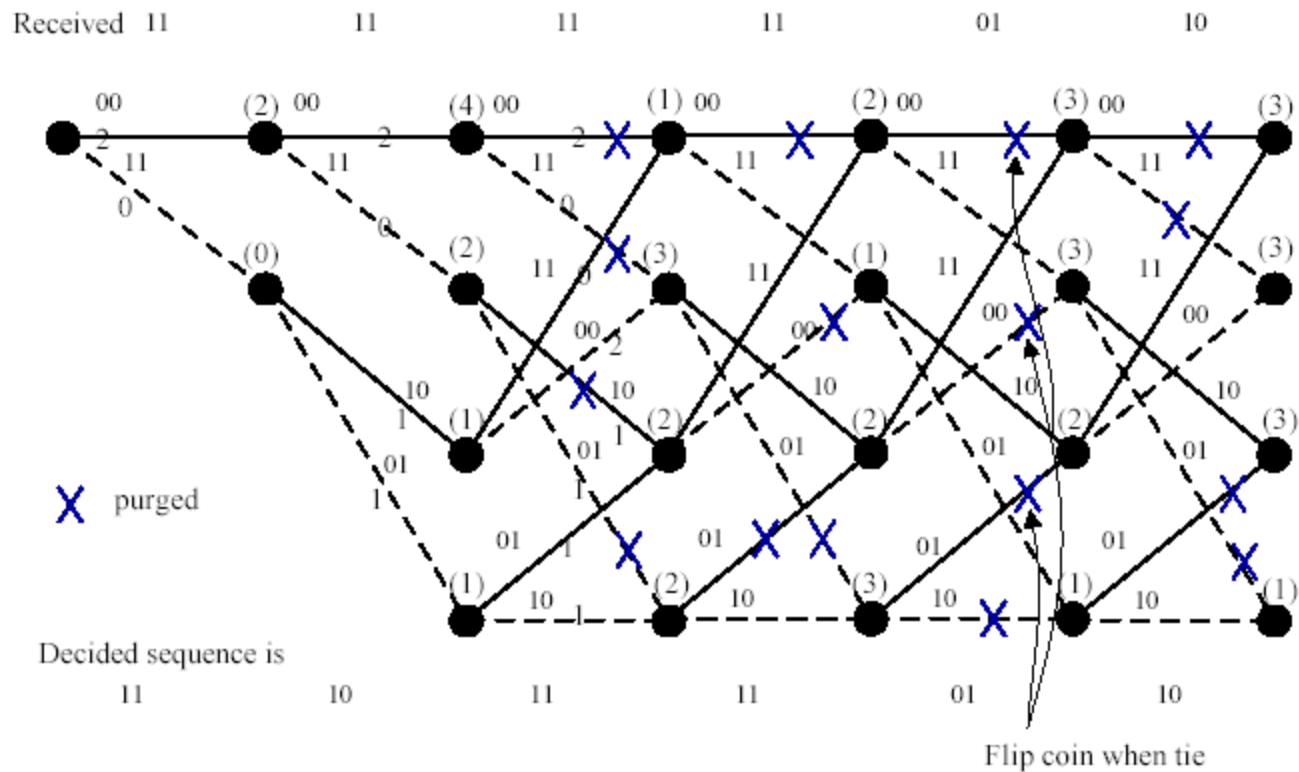
Trellis diagrams
Viterbi algorithm



Trellis diagram



Trellis decoding



Viterbi Algorithm

- Note that the number of states is 2^{memory}
- The number of branches is $2^{(\text{memory}+1)}$
- Let $\sigma(j,k)$ be the partial cumulative metric at state j at k -th trellis section
 1. Set $\sigma(j=a,0)=0$ (Starting at the state a)
 2. At time k , compute the partial cumulative metrics for all paths merging to each state
 3. Set $\sigma(j, k)$ equal to the best partial path metric entering the node corresponding to state- j at time t . Break tie with coin flipping. Mark the the best metric path
 4. At the end of sequence, trace back the marked path for decoding

Bit rate calculations

$$\text{Useful data bit rate} = (B \times C \times M \times N) / T$$

B efficiency of the RS block (188/204 = 0.92)

C convolution code rate (1/2, 2/3, 3/4 etc)

M number of bits per carrier (2, 4, 6)

N number of carriers used (1512 for 2k, 6048 for 8k)

T total symbol duration time (894 us + 28 us for 8k and 1/32 guard, 224 us + 7us for 2k mode and 1/32 guard)